



# BIOE50011 – Signals and <u>Control</u>

### MATLAB practical 2 – Poles and Stability

12 December, 2022

### Learning objective

By the end of this MATLAB session you should be able to:

- Understand and derive the characteristics of electric circuit components: resistor *R*, capacitor *C*, inductor *L*
- Solve linear dynamic equations using the Laplace transform
- Determine how the system's dynamics depend on the poles of the transfer function
- Explain how the system components affect the poles in the *s*-plane

## Poles and Stability – Recap

For a LTI system  $G(s) = \frac{N(s)}{D(s)}$ Poles,  $p_i$  are the solution of D(s) = 0

#### Three types of stability

- Asymptotic stability:  $Re(p_i) < 0$
- Instability:
  - $Re(p_i) > 0$
- Marginal stability:  $Re(p_i) = 0$



### Task 2 - RLC circuit

Different values of *R*, *L* and *C* result in different poles in the *s*-plane and influence the systems dynamics





### Task 2-Step responses and poles in the s-plane



## Task 3 - Closed-loop control of the wheel system



$$G(s) = \frac{10^5}{5s^3 + 5001s^2 + 2000s + 10^5K}$$
  
final value theorem  
$$\lim_{s \to 0} s \frac{1}{s} G(s) = \frac{1}{K}$$

#### When K = 1

- all poles have a <u>negative real</u> part
- the step response converges to  $\frac{1}{K} = 1$

# Task 3 - Step response and poles for varied K

### Step response and stability

### > K < 20:asymptotically stable

- Dominant poles have negative real parts
- Oscillation, converging to 1/K

### > K = 20: Marginal stable

- Real part of the dominant poles is 0
- Undamped oscillation

### > K > 20: unstable

- Dominant poles have positive real parts
- Oscillation with an increasing amplitude



### Task 3 - root locus



$$G(s) = \frac{10^5}{5s^3 + 5001s^2 + 2000s + 10^5K}$$

- A *root locus* plot is a graphical representation of the **poles' trajectory** in the complex plane with respect to the **variation of the feedback** <u>gain</u> (*K*).
- Poles are labelled by x, zeros by o (if there is any)
- If we add poles with the variation of K on root locus plot, poles will overlay on the root locus trajectories!

Trajectory 2: locations of the second dominant pole with a varying *K* 

### Task 4 – Performance of an approximated system



(1) Poles of the fan system

$$G_{\omega}(s) = \frac{10^5}{5s^2 + 5001s + 2000}$$

Poles 
$$p_{1,2} = \frac{-5001 \pm \sqrt{5001^2 - 4 \cdot 5 \cdot 2000}}{2 \cdot 5} \Rightarrow p_1 = -0.4 \text{ and } p_2 = -999.8$$

#### negative real poles → stable!

(2) DC gain of the fan system  $G_{\omega}(s)$  $\alpha = \lim_{s \to 0} s\left(\frac{1}{s}G_{\omega}(s)\right) = \lim_{s \to 0} \frac{10^5}{5s^2 + 5001s + 2000} = 50$ 

The angular velocity of the fan will converge to 50 with the unit step input!

## Task 4 – Performance of an approximated system

#### (3)-(4) Dominant pole approximation

Dominant pole  $p = 0.4 \Rightarrow$ the approximated system (derived in SG 1)

$$\widehat{G_{\omega}}(s) = \frac{K}{s+0.4}$$

Step response via finite value theorem:  $\widehat{\omega}(t) = 50(u(t) - e^{-0.4t})$ 

As  $t \to \infty$ , the step response of the system converges to 50 rad/s.

- The fan will converge to a fixed wheel angular speed after a period time.
- We can control the angular speed of the fan!



The dominant pole is far away from another pole



<sup>➔</sup> good approximation!

## Task 4 – Performance of an approximated system





Can we control the wheel angle of a **prize wheel**?



- (5) Perform the same analysis to the approximated **prize wheel** system:
  - The step response does NOT converge.
  - As  $t \to \infty$ , the wheel angle tends to infinite large.
  - Therefore, the angle of the prize wheel cannot be controlled by a step input.

